

The University of Texas at Austin  
Dept. of Electrical and Computer Engineering  
Midterm #2

Date: April 8, 2009

Course: EE 313 Evans

Name: Set, Solution  
Last, First

- The exam is scheduled to last 50 minutes.
- Open books and open notes. You may refer to your homework assignments and the homework solution sets.
- Calculators are allowed.
- You may use any standalone computer system, i.e. one that is not connected to a network. ***Please disable all wireless connections on your computer system.***
- Please turn off all cell phones, pagers, and personal digital assistants (PDAs).
- All work should be performed on the quiz itself. If more space is needed, then use the backs of the pages.
- **Fully justify your answers.**

Problem	Point Value	Your score	Topic
1	20		Difference Equation
2	20		Discrete-Time Convolution
3	40		Transfer Functions
4	20		Potpourri
Total	100		

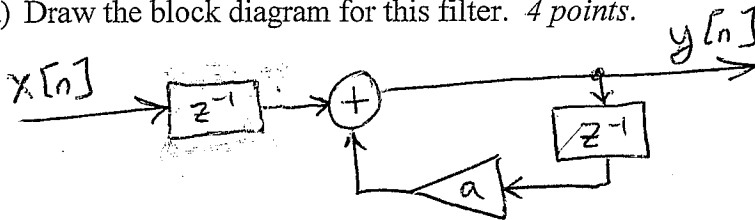
**Problem 2.1** Difference Equation. 20 points.

A causal discrete-time linear time-invariant filter with input  $x[n]$  and output  $y[n]$  is governed by the following difference equation

$$y[n] = a y[n-1] + x[n-1]$$

where  $|a| < 1$ .

- (a) Draw the block diagram for this filter. 4 points.



$z^{-1}$  represents a delay of one sample (slide 15-4). Lathi's book uses D operator instead of  $z^{-1}$ . (page 261).

- (b) What are the initial conditions? What values should they be assigned and why? 4 points.

Initial conditions can be found by setting  $n=0$  and advancing  $n$ :

$$y[0] = a y[-1] + x[-1] \Rightarrow \text{Initial conditions are } y[-1] \text{ and } x[-1].$$

$$y[1] = a y[0] + x[0]$$

The initial conditions should be set to zero for the system to be LTI.

- (c) What is (are) the values of the characteristic root(s)? 4 points.

The characteristic polynomial is

$$1 - a\gamma^{-1} = 0$$

which gives one characteristic root at

$$\gamma = a$$

- (d) Is the system bounded-input bounded-output stable? Why or why not? 4 points.

Since  $|a| < 1$ , the characteristic root is inside the unit circle. Zero-input solution is asymptotically stable (Lathi, p. 314), which implies BIBO stability (Lathi, p. 314).

- (e) Let the input  $x[n] = \delta[n]$ . Compute  $y[0]$ ,  $y[1]$  and  $y[2]$ . Infer a formula for the impulse response. 4 points.

$$y[0] = a y[-1] + \delta[-1] = 0$$

$$y[1] = a y[0] + \delta[0] = 1$$

$$y[2] = a y[1] + \delta[1] = a$$

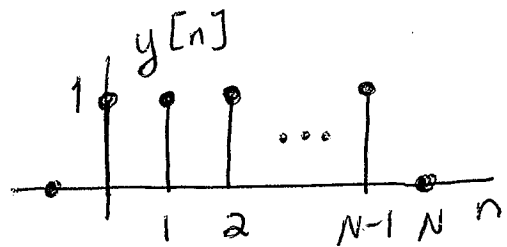
Here,  $y[0]$ ,  $y[1]$  and  $y[2]$  are first three samples of the impulse response.  $h[n] = a^{n-1} u[n-1]$

**Problem 2.2** Discrete-Time Convolution. 20 points.

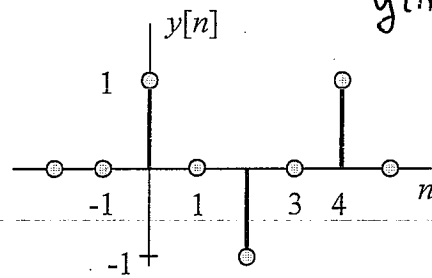
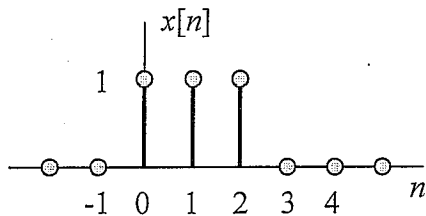
- (a) In discrete time, convolve the unit step function  $u[n]$  and the function  $\delta[n] - \delta[n-N]$ , where  $\delta[n]$  is the Kronecker impulse and  $N$  is a positive integer. 10 points.

$$\begin{aligned} y[n] &= u[n] * (\delta[n] - \delta[n-N]) \\ &= u[n] * \delta[n] - u[n] * \delta[n-N] \\ &= u[n] - u[n-N] \end{aligned}$$

$y[n]$  is a causal rectangular pulse of length  $N$  samples.



- (b) Consider a discrete-time linear time-invariant system. For input  $x[n]$  given below, the system gives output  $y[n]$  below. What is the impulse response of the system? 10 points.



$$\begin{aligned} y[n] &= h[n] * x[n] \\ &= \sum_{k=-\infty}^{\infty} h[k] x[n-k] \end{aligned}$$

Since  $x[n]$  and  $y[n]$  are causal,  $h[n]$  must be causal.

Since  $x[n]$  and  $y[n]$  are finite in extent, assume that  $h[n]$  is finite in extent:

$$\begin{aligned} L_x + L_h - 1 &= L_y \Rightarrow L_h = L_y - L_x + 1 \\ L_h &= 5 - 3 + 1 = 3 \end{aligned}$$

Assume  $h[n] = h[0]\delta[n] + h_1\delta[n-1] + h_2\delta[n-2]$ .

$$y[0] = h[0]x[0] \Rightarrow h[0] = 1$$

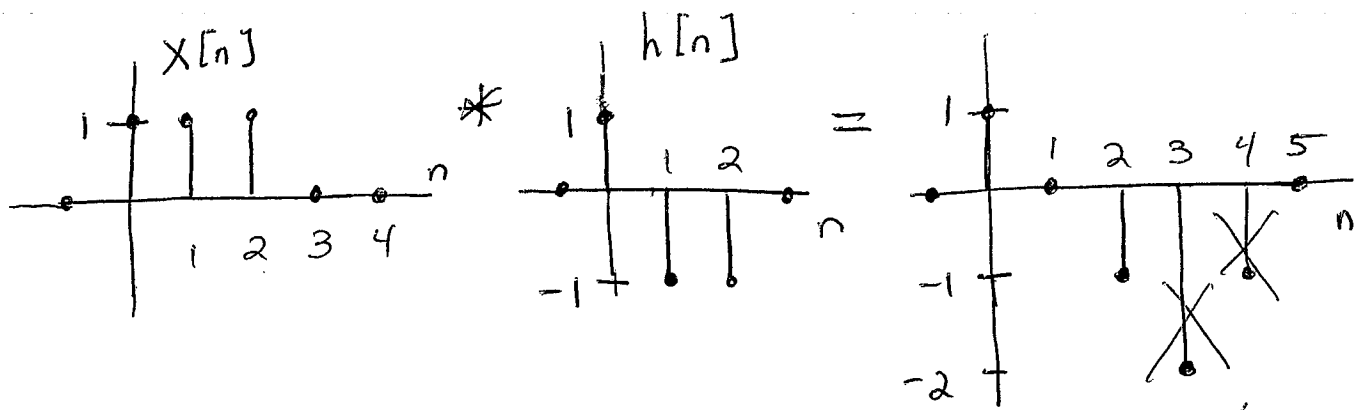
$$y[1] = h[0]x[1] + h[1]x[0] \Rightarrow h[1] = -1$$

$$y[2] = h[0]x[2] + h[1]x[1] + h[2]x[0]$$

$$-1 = 1 \cdot 1 + (-1) \cdot 1 + h[2] \cdot 1 \Rightarrow h[2] = -1$$

see next page  $\rightarrow$  L-39

2.2.(b) Continued.



The convolution produces the correct first three values of  $y[n]$  but is incorrect in the fourth and fifth values. Let's keep computing additional values of  $h[n]$ :

$$y[3] = h[0]x[3] + h[1]x[2] + h[2]x[1] + h[3]x[0] = 0$$

$$= 0 + (-1) + (-1) + h[3] = 0$$

$$h[3] = 2$$

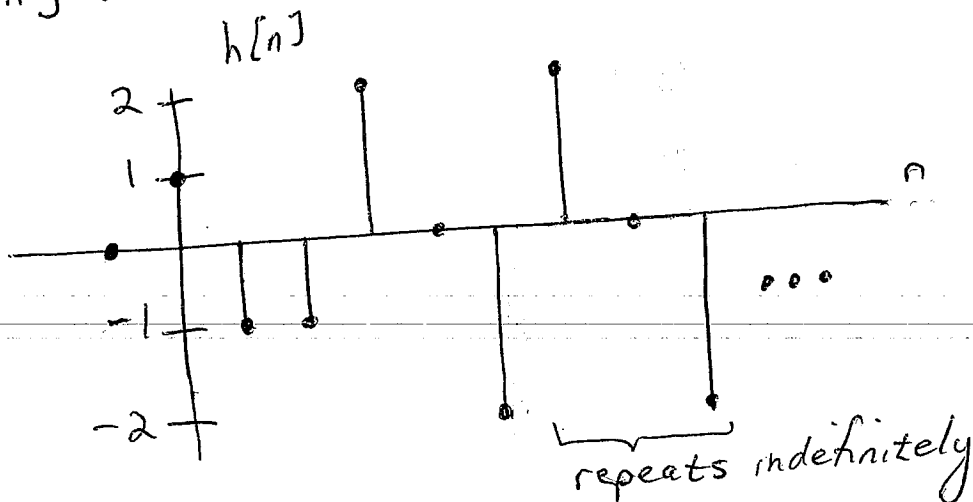
$$y[4] = h[2]x[2] + h[3]x[1] + h[4]x[0]$$

$$= (-1) + (2) + h[4] = 1$$

$$h[4] = 0$$

$$h[5] = -2$$

$h[n]$  is infinite in extent.



**Problem 2.3** Transfer Functions. 40 points.

A causal linear time-invariant (LTI) continuous-time system has the following transfer function in the Laplace transform domain:

$$H(s) = \frac{s}{s+1}$$

- (a) Find the corresponding differential equation using  $x(t)$  to denote the input signal and  $y(t)$  to denote the output signal. Give the minimum number of initial conditions, and their values. 8 points.

$$H(s) = \frac{s}{s+1} = \frac{Y(s)}{X(s)} \Rightarrow (s+1)Y(s) = sX(s) \Rightarrow sY(s) + Y(s) = sX(s)$$

Taking the inverse Laplace transform of both sides,  
 $y'(t) + y(t) = x'(t)$  with initial conditions

- (b) Give the pole location(s) and the region of convergence. 8 points.

Poles are the roots of the denominator of the transfer function, i.e.  $s = -1$ .

Region of convergence cannot include a pole location. Due to causality,

- (c) Compute the impulse response by taking the inverse Laplace transform of  $H(s)$ . 8 points.  $\text{Re}\{s\} > -1$

$$h(t) = \mathcal{L}^{-1}\{H(s)\} = \mathcal{L}^{-1}\left\{\frac{s}{s+1}\right\}$$

Solution #1: Partial fractions.

$$\mathcal{L}^{-1}\left\{\frac{s}{s+1}\right\} = \mathcal{L}^{-1}\left\{1 - \frac{1}{s+1}\right\} = \delta(t) - e^{-t}u(t)$$

Solution #2:  
 Differentiation Property

$$\mathcal{L}^{-1}\left\{\frac{1}{s+1}\right\} = e^{-t}u(t)$$

$$sF(s) \xrightarrow{\mathcal{L}^{-1}} \frac{d}{dt}f(t)$$

- (d) Give a formula for the frequency response. 8 points.

Since the region of convergence includes the imaginary axis,

$$H_{\text{freq}}(f) = H(s) \Big|_{s=j2\pi f} = \frac{j2\pi f}{j2\pi f + 1}$$

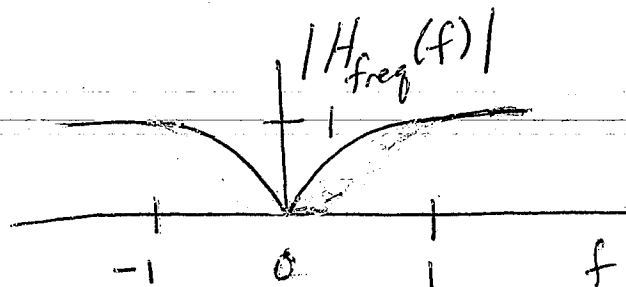
$$h(t) = -e^{-t}u(t) + e^{-t}\delta(t)$$

- (e) Plot the magnitude of the frequency response and describe the system's frequency selectivity (lowpass, highpass, bandpass, or bandstop). 8 points.

$$H_{\text{freq}}(0) = 0$$

$$\lim_{f \rightarrow \infty} |H_{\text{freq}}(f)| = \lim_{f \rightarrow \infty} \left| \frac{1}{1 + \frac{1}{j2\pi f}} \right| = 1$$

Highpass filter



**Problem 2.4 Potpourri.** 20 points.

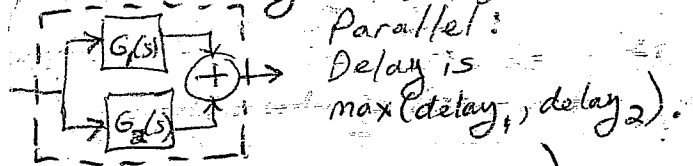
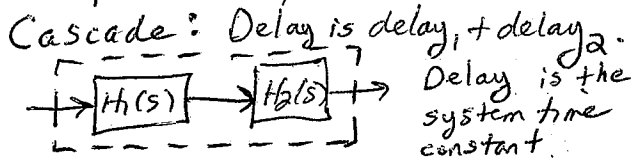
- (a) Either prove the following statement to be true, or give a counterexample to show that the following statement is false: The convolution of two continuous-time signals  $x(t)$  and  $y(t)$  may always be computed by taking the Laplace transforms of  $x(t)$  and  $y(t)$ , multiplying the Laplace transforms, and applying the inverse Laplace transform to the result. 5 points.

FALSE. Consider  $x(t) = e^{t^2} u(t)$ , which does not have a Laplace transform. See slide 11-11.

Also see page 405 of Lathi's book, second paragraph from bottom of page.

- (b) Consider wanting to map a transfer function in the Laplace domain into an implementation. The transfer function is a ratio of two polynomials. When would you recommend using a parallel implementation over a cascaded implementation? 5 points.

The parallel implementation would have shorter delay through system.



- (c) Give one application that uses a differentiator in continuous time and give one application that uses a differentiator in discrete time. 5 points.

A differentiator is a highpass filter - see slides 8-9 and 13-13 (in-class comments).

Discrete-time - detect edges and texture in images (slide 8-9). Enhance high frequencies in sampled audio (pp. 263-265) in Lathi.

The parallel implementation is a natural mapping of the result of applying partial fractions decomposition. A parallel form would be natural when numerator order is greater than denominator order.

- (d) Given a system governed by a linear constant-coefficient differential equation, the zero-state solution is always bounded-input bounded-output unstable if the zero-input solution is unstable. 5 points. Either prove the preceding statement to be true, or

Continuous-time - estimate acceleration from a velocity signal. Differential amplifiers.

give a counterexample to show that the preceding statement is false. FALSE.

Consider slide 13-10, which is from example 4.14 in Lathi.

Here is a simpler example based example 4.14:

$$y'(t) - y(t) = x'(t) - x(t)$$

Zero-input solution is of the form  $C_0 e^t u(t)$ . Unstable.

Zero-state solution is  $h(t) * x(t)$   $H(s) = \frac{s-1}{s-1} = 1$ .

Hence,  $h(t) = \delta(t)$ . Zero-state solution is simply  $x(t)$ .

Bounded input gives bounded zero-state solution.